

OP-AMP ve FİLTRELER

Prof.Dr.Remzi YILDIRIM

2016-AYBÜ-ANKARA

Giriş ve geri-besleme çevrimlerinde bir çok gerçekleştirilmesi mümkün transfer çevrimi kullanılabilir. (Tablo 2.4) de, uygulamada en çok kullanılan haller bir akım, şeması takib e ilerek verilmiştir.

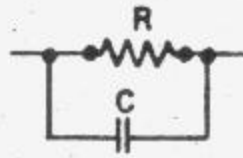
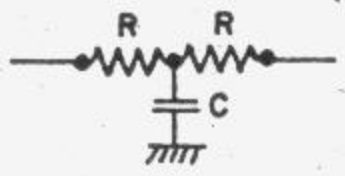
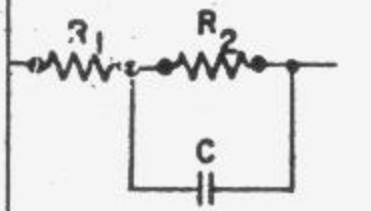
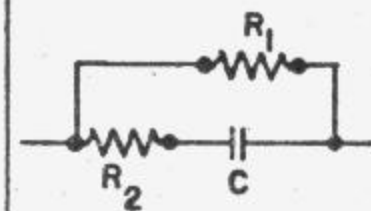
(Tablo 2.5) de, işlem yükselteçlerinin giriş ve geri-besleme çevrimlerinde kullanılabilecek genelleştirilmiş transfer empedansları verilmiştir. Bu çevrimler ve tekabül ettikleri transfer-empedans fonksiyonlarının oranlanması ile istenilen karmaşık işlemler gerçekleştirilebilir. (Tablo 2.5) in bağıntılar sütununda transfer empedans katsayıları, çevrim parametrelerinin fonksiyonu olarak verilmiştir. Ters bağıntılar ise çevrim parametrelerini transfer empedansı katsayıları cinsinden vermektedir. (Tablo 2.5) in nasıl kullanıldığının bir örneğini vermek üzere, (Tablo 2.4) ün (10.) işlem çevriminin transfer fonksiyonu (2.5) tablosu yardımıyla elde edilmesini göz önüne alalım. Burada geri-besleme transfer empedansı çevrimi:

TABLO 2.4

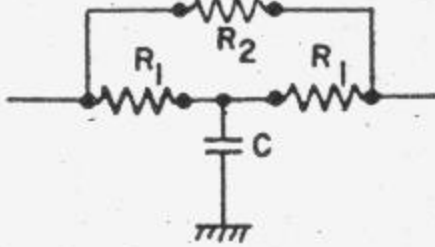
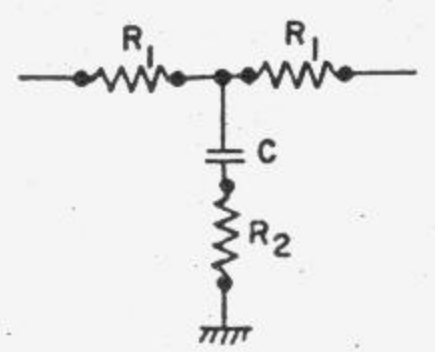
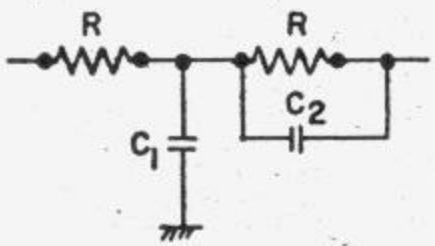
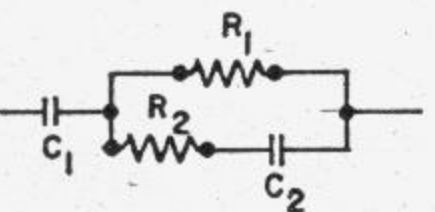
TRANSFER FONKSİYONU $\bar{e}_o / \bar{e}_i$	İŞLEM ÇEVİRİMİ
-1	①
$-\frac{1}{sRC}$	②
-sRC	③
$-\frac{R_2}{R_1} \left[\frac{1}{1+sR_2C_2} \right]$	④

TABLO 2.4 (Devam)

$-\frac{R_2}{R_1} [1+sR_1C_1]$	⑤
$-\left[\frac{R_2}{R_1} + \frac{1}{sR_1C_2} \right]$	⑥
$-\left[\frac{R_2}{R_1} \cdot \frac{1+sC_1R_1}{1+sC_2R_2} \right]$	⑦
$-\left[\frac{SC_1}{SC_2} \cdot \frac{1+sR_2C_2}{1+sR_1C_1} \right]$	⑧
$-\frac{(1+sR_1C_1) \cdot (1+sR_2C_2)}{sC_2R_1}$	⑨
$-\frac{s \cdot R_2C_1}{(1+sR_1C_1)(1+sR_2C_2)}$	⑩

No.	Transfer Empedansı Fonksiyonu	Çevrim	Bağıntılar	Ters Bağlantılar
1	A		$A = R$	$R = A$
2	$\frac{A}{\tau s + 1}$ $\frac{R \cdot X_C}{R + X_C}$		$A = R$ $\tau = RC$	$R = A$ $C = \frac{\tau}{A}$
3	$A(\tau s + 1)$		$A = 2R$ $\tau = \frac{RC}{2}$	$R = \frac{A}{2}$ $C = \frac{4\tau}{A}$
4	$A \left \frac{\theta \tau s + 1}{\tau s + 1} \right $ $\theta < 1$		$A = R_1 + R_2$ $\tau = R_2 C$ $\theta = \frac{R_2}{R_1 + R_2}$	$R_1 = A\theta$ $R_2 = A(1 - \theta)$ $C = \frac{\tau}{A(1 - \theta)}$
			$A = R_1$ $\tau = (R_1 + R_2)C$ $\theta = \frac{R_2}{R_1 + R_2}$	$R_1 = A$ $R_2 = \frac{A\theta}{1 - \theta}$ $C = \frac{\tau(1 - \theta)}{A}$

TABLO 2.5 (Devam)

5	$A \left \frac{\tau s + 1}{\theta \tau s + 1} \right $ $\theta < 1$		$A = \frac{2R_1 R_2}{2R_1 + R_2}$ $\tau = \frac{R_1 C}{2}$ $\theta = \frac{2R_1}{2R_1 + R_2}$	$R_1 = \frac{A}{2(1 - \theta)}$ $R_2 = \frac{A}{\theta}$ $C = \frac{4\tau(1 - \theta)}{A}$
			$A = 2R$ $\tau = \left R_2 + \frac{R_1}{2} \right C$ $\theta = \frac{2R_2}{2R_2 + R_1}$	$R_1 = \frac{A}{2}$ $R_2 = \frac{A\theta}{4(1 - \theta)}$ $C = \frac{4\tau(1 - \theta)}{A}$
			$A = 2R$ $\tau = \frac{R}{2} (C_1 + C_2)$ $\theta = \frac{2C_2}{C_1 + C_2}$	$R = \frac{A}{2}$ $C_1 = \frac{2\tau(2 - \theta)}{A}$ $C_2 = \frac{2\tau\theta}{A}$
6	$\frac{1}{Bs} \left \frac{(\tau_{1s} + 1)(\tau_{3s} + 1)}{\tau_{2s} + 1} \right $ $\tau_1 < \tau_2 < \tau_3$		$B = C_1$ $\tau_2 = (R_1 + R_2)C_2$	$R_1 = \frac{\tau_1 + \tau_3 - \tau_2}{B}$ $R_2 = \frac{\tau_1 \tau_3 (\tau_1 + \tau_3 - \tau_2)}{B(\tau_3 - \tau_2)(\tau_2 - \tau_1)}$

$$X_C = \frac{1}{2\pi f C}$$

$$\left(R \cdot \frac{1}{2\pi f C} \right) 2\pi f C$$

$$\left(R + \frac{1}{2\pi f C} \right) 2\pi f C$$

$$2\pi f C + 1$$

$$\frac{R}{1 + 2\pi f RC}$$

$$\frac{R}{s\tau}$$

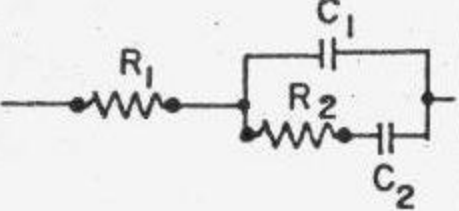
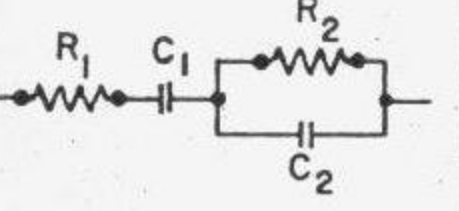
$$R \rightarrow A$$

$$\tau s + 1$$

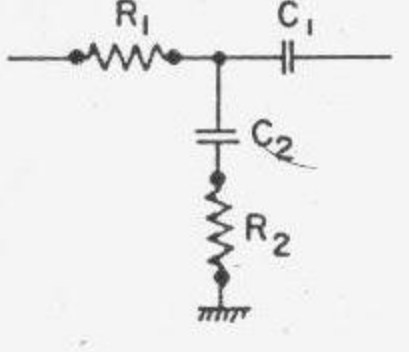
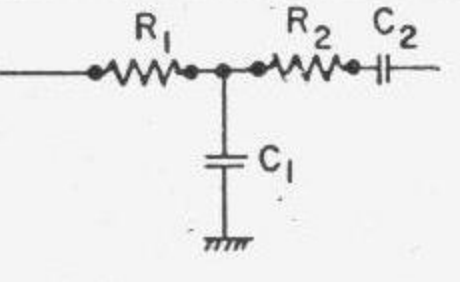
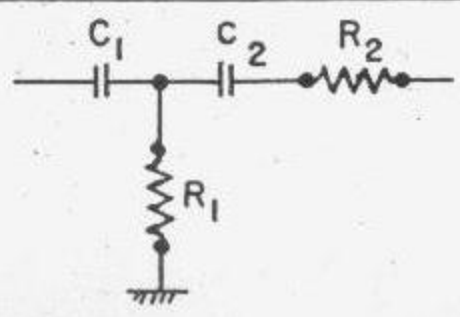
TABLO 2.5

15

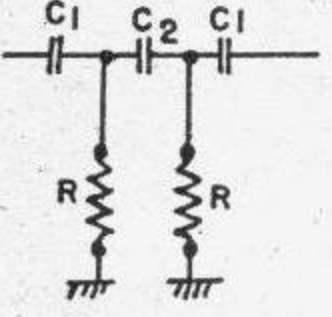
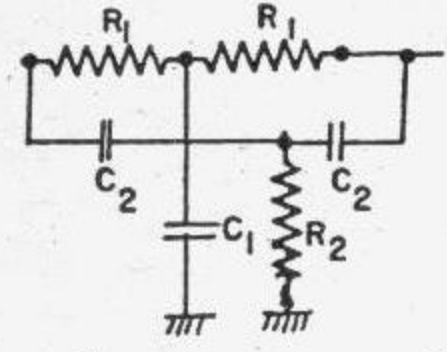
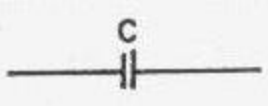
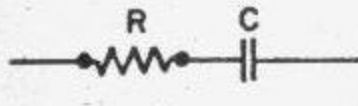
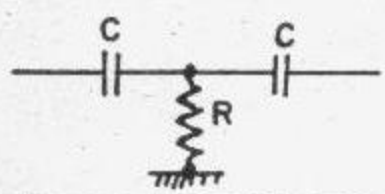
TABLO 2.5 (Devam)

6	$\frac{1}{Bs} \left \frac{(\tau_1 s + 1)(\tau_3 s + 1)}{\tau_2 s + 1} \right $ $\tau_1 < \tau_2 < \tau_3$		$\tau_1 \tau_3 = R_1 R_2 C_1 C_2$ $\tau_1 + \tau_3 = R_1 C_1 + R_2 C_2 + R_1 C_2$	$C_1 = B$ $C_2 = \frac{B(\tau_3 - \tau_2)(\tau_2 - \tau_1)}{(\tau_1 + \tau_3 - \tau_2)^2}$
			$B = C_1$ $\tau_2 = R_2 C_2$ $\tau_1 \tau_3 = R_1 R_2 C_1 C_2$ $\tau_1 + \tau_3 = R_1 C_1 + R_2 C_2 + R_2 C_1$	$R_1 = \frac{\tau_1 + \tau_3}{B \tau_2}$ $R_2 = \frac{(\tau_1 \tau_2 + \tau_2 \tau_3 - \tau_1 \tau_3)^2}{B \tau_2 (\tau_3 - \tau_2)(\tau_2 - \tau_1)}$ $C_1 = \frac{B \tau_2^2}{\tau_1 + \tau_2 + \tau_2 \tau_3 - \tau_1 \tau_3}$ $C_2 = \frac{B(\tau_3 - \tau_2)(\tau_2 - \tau_1)}{\tau_1 \tau_2 + \tau_2 \tau_3 - \tau_1 \tau_3}$

TABLO 2.5 (Devam)

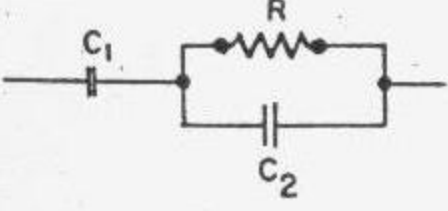
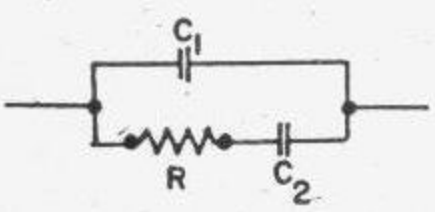
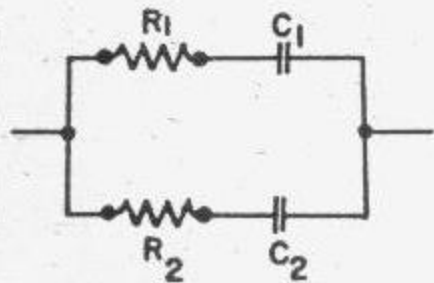
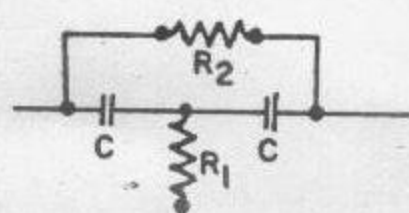
6	$\frac{1}{Bs} \left \frac{(\tau_1 s + 1)(\tau_3 s + 1)}{\tau_2 s + 1} \right $ $\tau_1 < \tau_2 < \tau_3$		$B = C_1$ $\tau_2 = R_2 C_2$ $\tau_1 \tau_3 = R_1 R_2 C_1 C_2$ $\tau_1 + \tau_3 = R_1 C_1 + R_2 C_2 + R_1 C_2$	$R_1 = \frac{\tau_1 \tau_3}{B \tau_2}$ $R_2 = \frac{\tau_1 \tau_2 \tau_3}{B(\tau_3 - \tau_2)(\tau_2 - \tau_1)}$ $C_1 = B$ $C_2 = \frac{B(\tau_3 - \tau_2)(\tau_2 - \tau_1)}{\tau_1 \tau_3}$
7	$\frac{1}{Bs} (1 + \tau_1 s)(1 + \tau_2 s)$ $\tau_1 \neq \tau_2$		$B = C_2$ $\tau_1 \tau_2 = R_1 R_2 C_1 C_2$ $\tau_1 + \tau_2 = R_1 C_1 + R_2 C_2 + R_1 C_2$	$R_1 = \frac{(\sqrt{\tau_1} - \sqrt{\tau_2})^2}{B}$ $R_2 = \frac{\sqrt{\tau_1 \tau_2}}{B}$ $C_1 = \frac{B \sqrt{\tau_1 \tau_2}}{(\sqrt{\tau_1} - \sqrt{\tau_2})^2}$ $C_2 = B$
8	$\frac{1}{Bs} \left \frac{(1 + \tau_1 s)(1 + \tau_2 s)}{\sqrt{\tau_1 \tau_2} s} \right $ $\tau_1 \neq \tau_2$		$B = C$ $\tau_1 \tau_2 = R_1 R_2 C_1 C_2$ $\tau_1 + \tau_2 = R_1 C_1 + R_2 C_2 + R_1 C_2$	$R_1 = \frac{(\sqrt{\tau_1} - \sqrt{\tau_2})^2}{B}$ $R_2 = \frac{\sqrt{\tau_1 \tau_2}}{B}$ $C_1 = \frac{B \sqrt{\tau_1 \tau_2}}{(\sqrt{\tau_1} - \sqrt{\tau_2})^2}$ $C_2 = B$

TABLO 2.5 (Devam)

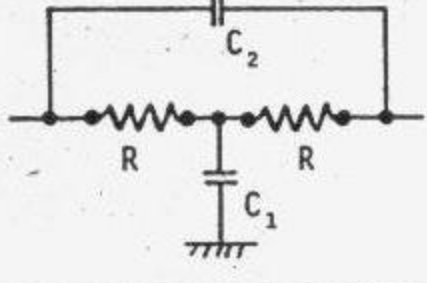
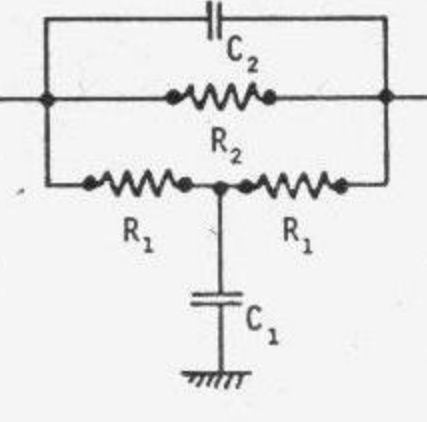
9	$\frac{1}{Bs} \left \frac{(\tau_1 s + 1)(\tau_2 s + 1)}{\tau_1 \tau_2 s^2} \right $ $\tau_1 < \tau_2$		$B = \frac{C_1 C_2}{C_1 + 2C_2}$ $\tau_1 = RC_1$ $\tau_2 = R(C_1 + 2C_2)$	$R = \frac{\tau_1(\tau_2 - \tau_1)}{2B\tau_2}$ $C_1 = \frac{2B\tau_2}{\tau_2 - \tau_1}$ $C_2 = \frac{B\tau_2}{\tau_1}$
10	$A \left \frac{\tau_1 s + 1}{\tau_1 \tau_2 s^2 + 1} \right $	 $R_1 C_1 = 4R_2 C_2$	$A = 2R_1$ $\tau_1 = \frac{R_1 C_1}{2} = 2R_2 C_2$ $\tau_2 = R_1 C_2$	$R_1 = \frac{A}{2}$ $R_2 = \frac{A\tau_1}{4\tau_2}$ $C_1 = \frac{4\tau_1}{A}$ $C_2 = \frac{2\tau_2}{A}$
11	$\frac{1}{Bs}$		$B = C$	$C = B$
12	$\frac{(\tau s + 1)}{Bs}$		$B = C$ $\tau = RC$	$R = \frac{\tau}{B}$ $C = B$
13	$\frac{1}{Bs} \left \frac{\tau s + 1}{\tau s} \right $		$B = \frac{C}{2}$ $\tau = 2RC$	$R = \frac{\tau}{4B}$ $C = 2B$

4

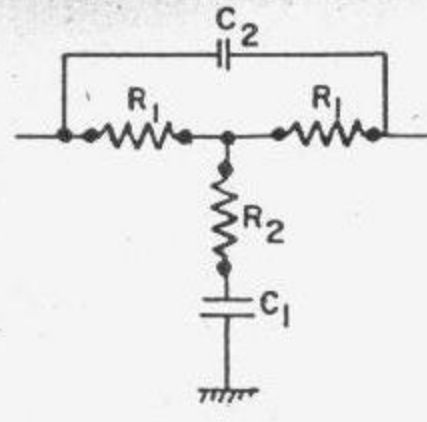
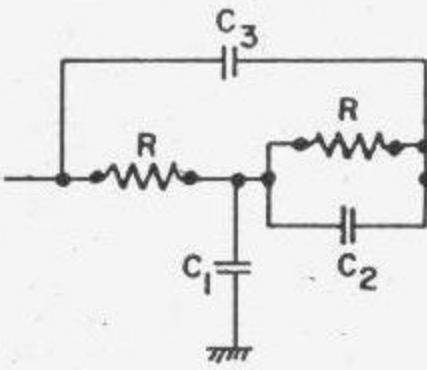
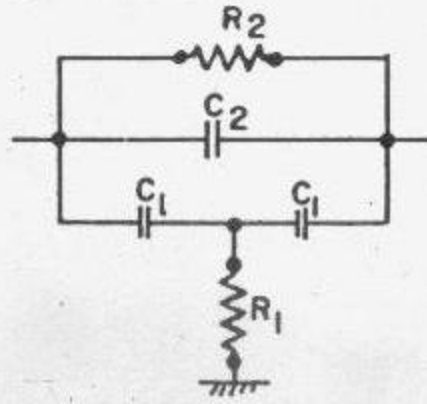
TABLO 2.5 (Devam)

14	$\frac{1}{Bs} \left \frac{\tau s + 1}{\tau s} \right $ $\theta < 1$		$B = C_1$ $\tau = \frac{R(C_1 + C_2)}{C_1 + C_2}$ $\theta = \frac{C_2}{C_1 + C_2}$	$R = \tau(1 - \theta)$ $C_1 = B$ $C_2 = \frac{B\theta}{1 - \theta}$
			$B = C_1 + C$ $\tau = RC_2$ $\theta = \frac{C_2}{C_1 + C_2}$	$R = \frac{\tau}{B(1 - \theta)}$ $C_1 = B\theta$ $C_2 = B(1 - \theta)$
15	$\frac{1}{Bs} \left \frac{(\tau_1 s + 1)(\tau_3 s + 1)}{\tau_2 s + 1} \right $ $\tau_1 < \tau_2 < \tau_3$		$B = C_1 + C_2$ $\tau_1 = R_1 C_1$ $\tau_2 = (R_1 + R_2) \left \frac{C_1 C_2}{C_1 + C_2} \right $ $\tau_3 = R_2 C_2$	$R_1 = \frac{\tau_1(\tau_3 - \tau_1)}{B(\tau_2 - \tau_1)}$ $R_2 = \frac{\tau_3(\tau_3 - \tau_1)}{B(\tau_3 - \tau_2)}$ $C_1 = \frac{B(\tau_2 - \tau_1)}{\tau_3 - \tau_1}$ $C_2 = \frac{B(\tau_3 - \tau_2)}{\tau_3 - \tau_1}$
16	$A \left \frac{\tau_1 s + 1}{\tau_1 \tau_2 s^2 + \tau_1 s + 1} \right $		$A = R_2$ $\tau_1 = 2R_1 C$ $R_2 C$	$R_1 = \frac{A\tau_1}{4\tau_2}$ $R_2 = A$ $2\tau_2$

TABLO 2.5 (Devam)

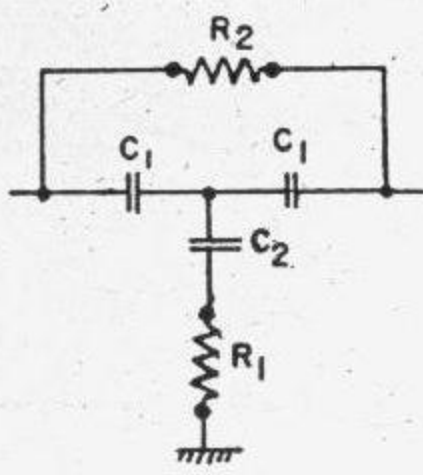
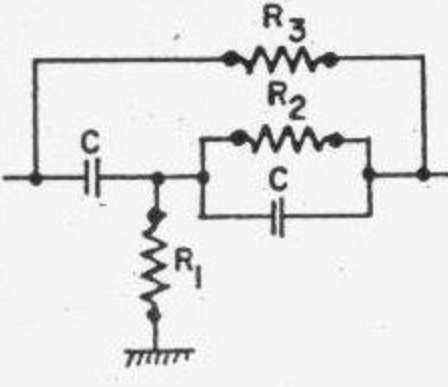
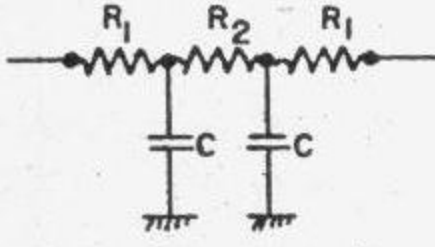
17	$A \cdot \left \frac{\tau_2 s + 1}{\tau_1 \tau_2 s^2 + \tau_1 s + 1} \right $		$A = 2R$ $\tau_1 = 2RC_2$ $\tau_2 = \frac{RC_1}{2}$	$R = \frac{A}{2}$ $C_1 = \frac{4\tau_2}{A}$ $C_2 = \frac{\tau_1}{A}$
18	$A \cdot \left \frac{\tau_3 s + 1}{\tau_1 \tau_2 s^2 + \tau_1 s + 1} \right $ $\tau_2 > \frac{\tau_1}{4}$ (kompleks kökler) $\tau_3 > \tau_2$		$A = \frac{2R_1 R_2}{(2R_1 + R_2)}$ $\tau_1 = \frac{R_1(R_1 C_1 + 2R_2 C_2)}{2R_1 + R_2}$ $\tau_2 = \frac{R_1 R_2 C_1 C_2}{R_1 C_1 + 2R_2 C_2}$ $\tau_3 = \frac{R_1 C_1}{2}$	$R_1 = \frac{A\tau_3^2}{2 \tau_3^2 - \tau_1(\tau_3 - \tau_2) }$ $R_2 = \frac{A\tau_3^2}{\tau_1(\tau_3 - \tau_2)}$ $C_1 = \frac{4 \tau_3^2 - \tau_1(\tau_3 - \tau_2) }{A\tau_3}$ $C_2 = \frac{\tau_1 \tau_2}{A\tau_3}$

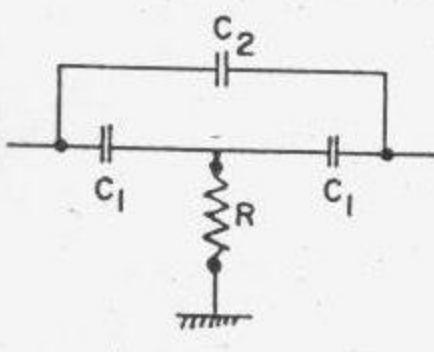
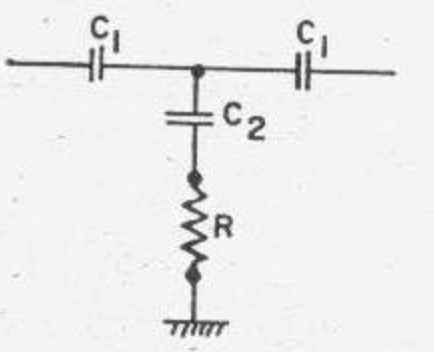
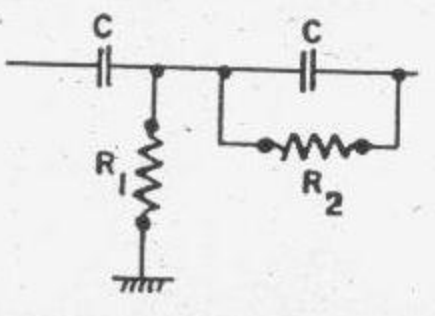
TABLO 2.5 (Devam)

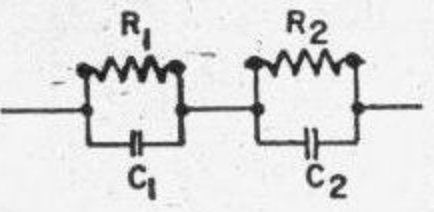
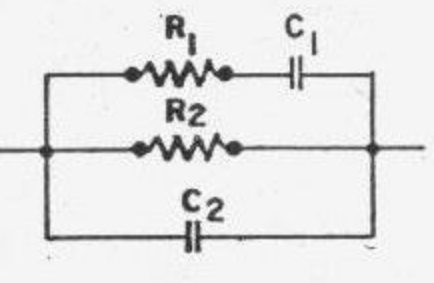
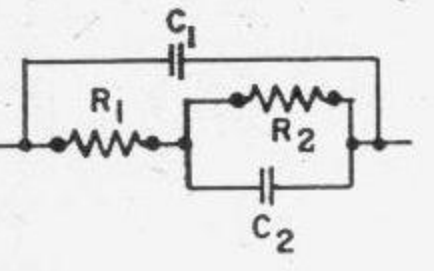
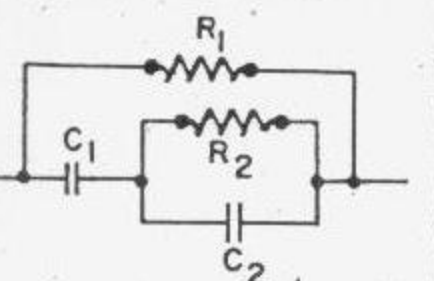
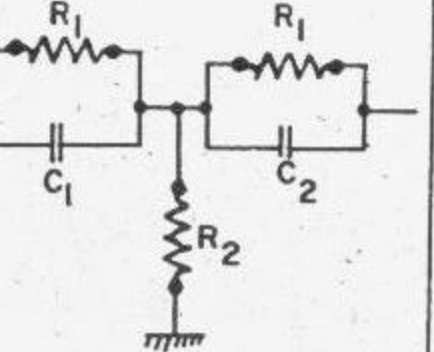
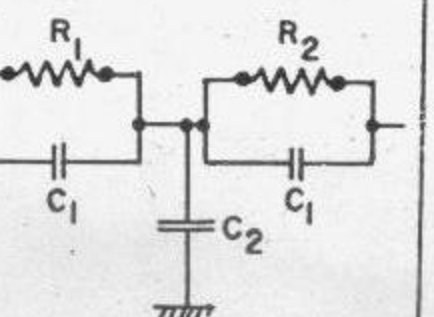
18	$A \cdot \left \frac{\tau_3 s + 1}{\tau_1 \tau_2 s^2 + \tau_1 s + 1} \right $ $\tau_2 > \frac{\tau_1}{4}$ (kompleks kökler) $\tau_3 > \tau_2$		$A = 2R_1$ $\tau_1 = R_2 C_1 + 2R_1 C_2$ $\tau_2 = \frac{R_1(R_1 + 2R_2)C_1 C_2}{R_2 C_1 + 2R_1 C_2}$ $\tau_3 = \left R_2 + \frac{R_1}{2} \right C_1$	$R_1 = \frac{A}{2}$ $R_2 = \frac{A\tau_1(\tau_3 - \tau_2)}{4 \tau_3^2 - \tau_1(\tau_3 - \tau_2) }$ $C_1 = \frac{4 \tau_3^2 - \tau_1(\tau_3 - \tau_2) }{A\tau_3}$ $C_2 = \frac{\tau_1 \tau_2}{A\tau_3}$
			$A = 2R$ $\tau_1 = R(C_2 + 2C_3)$ $\tau_2 = \frac{RC_3(C_1 + C_2)}{C_2 + 2C_3}$ $\tau_3 = \frac{R}{2}(C_1 + C_2)$	$R = \frac{A}{2}$ $C = \frac{2 2\tau_3^2 - \tau_1(\tau_3 - \tau_2) }{A\tau_3}$ $C_2 = \frac{2\tau_1(\tau_3 - \tau_2)}{A\tau_3}$ $C_3 = \frac{\tau_1 \tau_2}{A\tau_3}$
19	$A \cdot \left \frac{\tau_3 s + 1}{\tau_1 \tau_2 s^2 + \tau_1 s + 1} \right $ $\tau_2 > \frac{\tau_1}{4}$ (kompleks kökler) $\tau_3 < \tau_1$		$A = R_2$ $\tau_1 = 2R_1 C_1 + R_2 C_2$ $\tau_2 = \frac{R_1 R_2 C_1(C_1 + 2C_2)}{2R_1 C_1 + R_2 C_2}$ $\tau_3 = 2R_1 C_1$	$R_1 = \frac{A\tau_3^2}{4 \tau_1 \tau_2 - \tau_3(\tau_1 - \tau_3) }$ $R_2 = A$ $C = \frac{2 \tau_1 \tau_2 - \tau_3(\tau_1 - \tau_3) }{A\tau_3}$ $C_2 = \frac{\tau_1 - \tau_3}{A}$

Transfer Fonk

TABLO 2.5 (Devam)

19	$A \cdot \left \frac{\tau_3 s + 1}{\tau_1 \tau_2 s^2 + \tau_1 s + 1} \right $ $\tau_2 > \frac{\tau_1}{4}$ (kompleks kökler) $\tau_3 < \tau_1$		$A = R_2$ $\tau_1 = \frac{C_1(2R_1C_2 + R_2C_1)}{2C_1 + C_2}$ $\tau_2 = \frac{R_1R_2C_1C_2}{2R_1C_2 + R_2C_1}$ $\tau_3 = \frac{2R_1C_1C_2}{(2C_1 + C_2)}$	$R_1 = \frac{A\tau_3^2}{4 \tau_1\tau_2 - \tau_3(\tau_1 - \tau_3) }$ $R_2 = A$ $C_1 = \frac{2\tau_1\tau_2}{A\tau_3}$ $C_2 = \frac{4\tau_1\tau_2 \tau_1\tau_2 - \tau_3(\tau_1 - \tau_3) }{A\tau_3^2(\tau_1 - \tau_3)}$
			$A = R_3$ $\tau_1 = \frac{R_1(2R_2 + R_3)C}{R_1 + R_2}$ $\tau_2 = \frac{R_2R_3C}{(2R_2 + R_3)}$ $\tau_3 = \frac{2R_1R_2C}{(R_1 + R_2)}$	$R_1 = \frac{A\tau_3^2}{2 2\tau_1\tau_2 - \tau_3(\tau_1 - \tau_3) }$ $R_2 = \frac{A\tau_3}{2(\tau_1 - \tau_3)}$ $R_3 = A$ $C = \frac{2\tau_1\tau_2}{A\tau_3}$
20	$A(\tau_1 s + 1)(\tau_2 s + 1)$ $\tau_1 < \tau_2$		$A = 2R_1 + R_2$ $\tau_1 = \left \frac{R_1R_2}{2R_1 + R_2} \right C$ $\tau_2 = R_1C$	$R_1 = A \left \frac{\tau_2 - \tau_1}{2\tau_2} \right $ $R_2 = A \frac{\tau_1}{\tau_2^2}$ $C = \frac{2\tau_2^2}{A(\tau_2 - \tau_1)}$

21	$\frac{1}{Bs} \cdot \left \frac{\theta\tau s + 1}{\tau s + 1} \right $ $\theta < 1$		$B_2 = C_2$ $\tau = RC_1 \left \frac{2C_2 + C}{C_2} \right $ $\theta = \frac{2C_2}{2C_2 + C_1}$	$R = \frac{\tau\theta^2}{4B(1-\theta)}$ $C_1 = \frac{2B(1-\theta)}{\theta}$ $C_2 = B$
			$B = \frac{C_1^2}{2C_1 + C_2}$ $\tau = RC_2$ $\theta = \frac{2C_1}{2C_1 + C_2}$	$R = \frac{\tau\theta^2}{4B(1-\theta)}$ $C_1 = \frac{2B}{\theta}$ $C_2 = \frac{4B(1-\theta)}{\theta^2}$
			$B = \left \frac{R_1}{R_1 + R_2} \right C$ $\tau = R_2C$ $\theta = \frac{2R_1}{R_1 + R_2}$	$R_1 = \frac{\tau\theta^2}{2B(2-\theta)}$ $R_2 = \frac{\tau\theta}{2B}$ $C = \frac{2B}{\theta}$

22	$A \cdot \left \frac{\tau_2 s + 1}{(\tau_1 s + 1)(\tau_3 s + 1)} \right $ $\tau_1 < \tau_2 < \tau_3$		$A = R_1 + R_2$ $\tau_1 = R_1 C_1$ $\tau_2 = \left \frac{R_1 R_2}{R_1 R_2} \right (C_1 + C_2)$ $\tau_3 = R_2 C_2$	$R_1 = \frac{A(\tau_2 - \tau_1)}{\tau_3 - \tau_1}$ $R_2 = \frac{A(\tau_3 - \tau_2)}{\tau_3 - \tau_1}$ $C_1 = \frac{\tau_1(\tau_3 - \tau_1)}{A(\tau_2 - \tau_1)}$ $C_2 = \frac{\tau_3(\tau_3 - \tau_1)}{A(\tau_3 - \tau_2)}$
			$A = R_2$ $\tau_2 = R_1 C_1$ $\tau_1 \tau_3 = R_1 R_2 C_1 C_2$ $\tau_1 + \tau_3 = R_1 C_1 + R_2 C_2 + R_2 C_1$	$R_1 = \frac{A \tau_2^2}{(\tau_3 - \tau_2)(\tau_2 - \tau_1)}$ $R_2 = A$ $C_1 = \frac{(\tau_3 - \tau_2)(\tau_2 - \tau_1)}{A \tau_2}$ $C_2 = \frac{\tau_1 \tau_3}{\tau_2}$
			$A = R_1 + R_2$ $\tau_2 = \left \frac{R_1 R_2}{R_1 + R_2} \right C$ $\tau_1 \tau_3 = R_1 R_2 C_1 C_2$ $\tau_1 + \tau_3 = R_1 C_1 + R_2 C_2 + R_2 C_1$	$R_1 = \frac{A \tau_2^2}{\tau_1 \tau_2 + \tau_2 \tau_3 - \tau_1 \tau_3}$ $R_2 = \frac{A(\tau_3 - \tau_2)(\tau_2 - \tau_1)}{\tau_1 \tau_2 + \tau_2 \tau_3 - \tau_1 \tau_3}$ $C_1 = \frac{\tau_1 \tau_3}{A \tau_2}$ $C_2 = \frac{(\tau_1 \tau_2 + \tau_2 \tau_3 - \tau_1 \tau_3)^2}{A \tau_2 (\tau_3 - \tau_2)(\tau_2 - \tau_1)}$
22	$A \cdot \left \frac{\tau_2 s + 1}{(\tau_1 s + 1)(\tau_3 s + 1)} \right $ $\tau_1 < \tau_2 < \tau_3$		$A = R_1$ $\tau_2 = R_2 (C_1 + C_2)$ $\tau_1 \tau_3 = R_1 R_2 C_1 C_2$ $\tau_1 + \tau_3 = R_1 C_1 + R_2 C_2 + R_2 C_1$	$R_1 = A$ $R_2 = \frac{A(\tau_3 - \tau_2)(\tau_2 - \tau_1)}{(\tau_1 + \tau_3 - \tau_2)^2}$ $C_1 = \frac{\tau_1 + \tau_3 - \tau_2}{A}$ $C_2 = \frac{\tau_1 \tau_3 (\tau_1 + \tau_3 - \tau_2)}{A(\tau_3 - \tau_2)(\tau_2 - \tau_1)}$
23	$A \cdot \left \frac{\tau_2 s + 1}{(\tau_1 s + 1)(\tau_3 s + 1)} \right $ $\tau_2 \leq \tau_1 \leq \tau_3$		$A = 2R + \frac{R_1^2}{R}$ $\tau_1 = R_1 C_1$ $\tau_2 = \left \frac{R_1 R_2}{R_1 + 2R_2} \right (C_1 + C_2)$ $\tau_3 = R_1 C_2$	$R_1 = \frac{A \tau_2}{(\tau_1 + \tau_3)}$ $R_2 = \frac{A \tau_2^2}{(\tau_1 + \tau_3)(\tau_1 + \tau_3 - 2\tau_2)}$ $C_1 = \frac{\tau_1(\tau_1 + \tau_3)}{A}$ $C_2 = \frac{\tau_3(\tau_1 + \tau_3)}{A \tau_2}$
24	$A \cdot \left \frac{\tau_2 s + 1}{(\tau_1 s + 1)(\tau_3 s + 1)} \right $ $\tau_1 \leq \tau_3 \leq \tau_2$		$A = R_1 + R_2$ $\tau_1 = R_1 C_1$ $\tau_2 = \frac{R_1 R_2}{R_1 + R_2} (2C_1 + C_2)$ $\tau_3 = R_2 C_1$	$R_1 = \frac{A \tau_1}{(\tau_1 + \tau_3)}$ $R_2 = \frac{A \tau_3}{(\tau_1 + \tau_3)}$ $C_1 = \frac{(\tau_1 + \tau_3)}{A}$ $C_2 = \frac{(\tau_1 + \tau_3)}{A} \left \frac{\tau_2}{\tau_3} + \frac{\tau_2}{\tau_1} - 2 \right $

The most troublesome faults are those that occur erratically. As a rule, they are caused by a loose soldered joint, less often by a broken resistor and only seldom by a fault inside a semiconductor component. In a functional model, pull the components gently and watch the suspected joint. When checking an evaluation sample, evoke the fault by twisting the printed-circuit board and watch (on an oscilloscope permanently connected to the output) whether the fault persists.

Faults, frequently occurring in manufacture are as follows: loose soldered joints, unsoldered joints, interrupted copper, solder bridge between two points, broken resistor, an order-of magnitude error in resistor value, interchanged resistors, back to front diode or electrolytic capacitor, interchanged pins of a transistor or operational amplifier, contact between two transistor cases, left out wire link.

Ac faults are relatively rare. If the operational circuit oscillates, the frequency compensations are usually not in order. The ceramic capacitor is a reliable component, its orientation does not matter, so that almost the only possible cause of failure is an incorrect value. If everything seems all right, re-solder all the soldered joints and check the series resistors. An oscillating power supply is a tricky fault to locate.

SUMMARY

Better than the best advice is one's own experience. Therefore, take an operational amplifier and build something on your own!

REFERENCES

- [1] J. I. Smith, Modern Operational Circuit Design, Wiley, New York, 1971, Chap. 15

SUMMARY TABLES

TABLE I

Ideal properties of basic operational circuits. For practical reasons, the voltage follower is shown explicitly

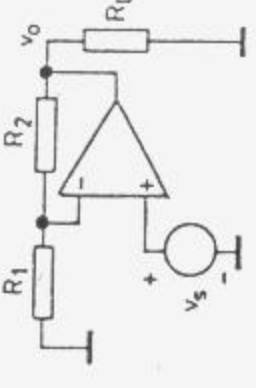
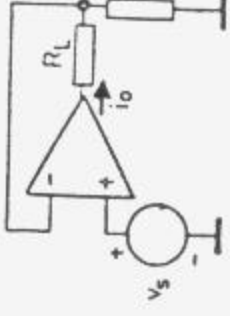
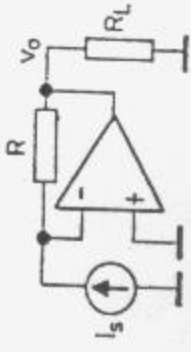
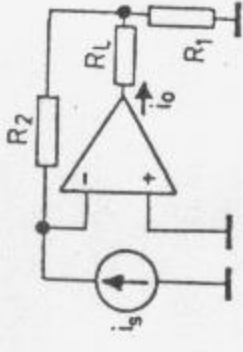
Operational circuit	Closed-loop gain G_1	Input resistance R_{i1}	Output resistance R_{o1}
A Voltage amplifier 	$\frac{R_2}{R_1} + 1$	∞	0
B V/I converter 	$\frac{1}{R}$	∞	∞
C I/V converter 	$-R$	0	0
D Current amplifier 	$-\left(\frac{R_2}{R_1} + 1\right)$	0	∞

TABLE I (continued)

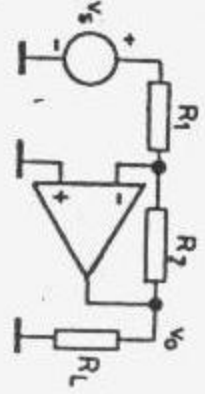
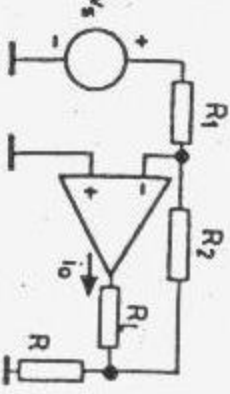
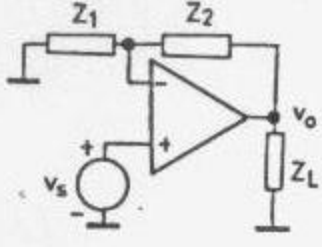
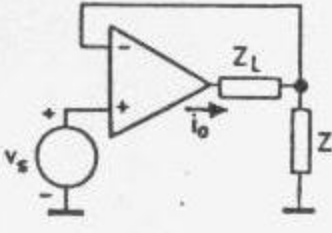
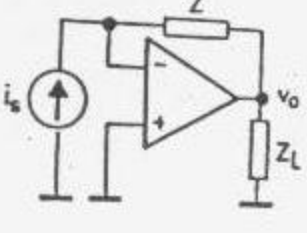
Operational circuit	Closed-loop gain G_1	Input resistance R_{i1}	Output resistance R_{o1}
E Voltage inverter 	$-\frac{R_2}{R_1}$	R_1	0
F Parallel V/I converter 	$-\frac{1}{R_1} \left(\frac{R_2}{R} + 1 \right)$	R_1	∞
G Voltage follower 	1	∞	0

TABLE II

Inverse feedback factor $1/\beta$ of basic operational circuits. Simplified expressions in the first column correspond to the ideal operational amplifier. All expressions make the assumption of an ideal signal source

Operational circuit		Inverse feedback factor $1/\beta$	
		ideal	actual
A		$1 + \frac{Z_2}{Z_1}$	$1 + \frac{Z_2}{Z_1 \parallel Z_d \parallel Z_{cm}^-} + \frac{Z_0}{Z_L} \left(1 + \frac{Z_2 + Z_L}{Z_1 \parallel Z_d \parallel Z_{cm}^-} \right)$
B		$1 + \frac{Z_L}{Z}$	$1 + \frac{Z_L + Z_0}{Z \parallel Z_d \parallel Z_{cm}^-}$
C		1	$1 + \frac{Z}{Z_d \parallel Z_{cm}^-} + \frac{Z_0}{Z_L} \left(1 + \frac{Z + Z_L}{Z_d \parallel Z_{cm}^-} \right)$

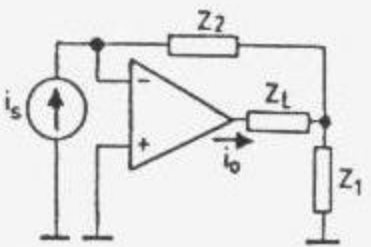
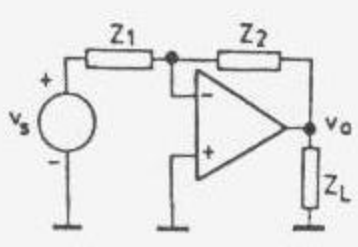
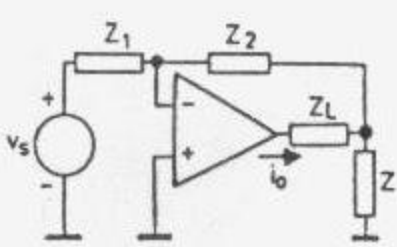
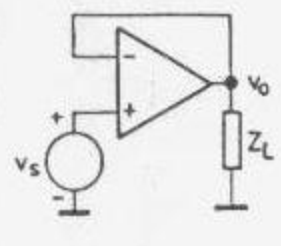
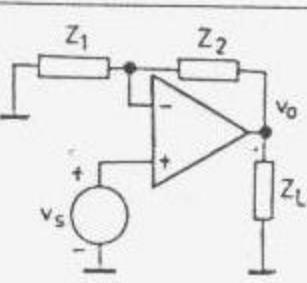
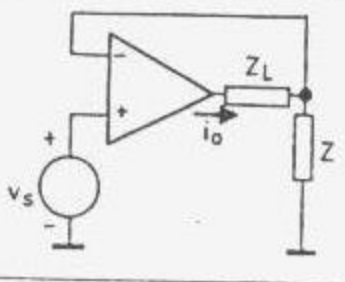
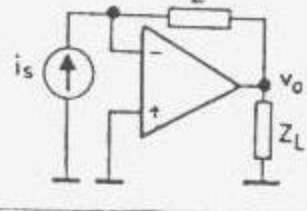
D		$1 + \frac{Z_L}{Z_1}$	$1 + \frac{Z_2}{Z_d \parallel Z_{cm}^-} + \frac{Z_L + Z_0}{Z_1} \left(1 + \frac{Z_2 + Z_1}{Z_d \parallel Z_{cm}^-} \right)$
E		$1 + \frac{Z_2}{Z_1}$	$1 + \frac{Z_2}{Z_1 \parallel Z_d \parallel Z_{cm}^-} + \frac{Z_0}{Z_L} \left(1 + \frac{Z_2 + Z_L}{Z_1 \parallel Z_d \parallel Z_{cm}^-} \right)$
F		$\left(1 + \frac{Z_2}{Z_1} \right) \left(1 + \frac{Z_L}{Z \parallel (Z_1 + Z_2)} \right)$	$1 + \frac{Z_2}{Z_1 \parallel Z_d \parallel Z_{cm}^-} + \frac{Z_L + Z_0}{Z} \left(1 + \frac{Z_2 + Z}{Z_1 \parallel Z_d \parallel Z_{cm}^-} \right)$
G		1	$1 + \frac{Z_0}{Z_L \parallel Z_d \parallel Z_{cm}^-}$

TABLE III

Components of closed-loop gain G of basic operational circuits,

$$G = \left(G_\infty + \frac{G_0}{\beta A} \right) \frac{\beta A}{1 + \beta A}$$

Expressions for the feedback factor β are in Tab. II

Operational circuit	Operational-circuit closed-loop gain		
	ideal G_1	idealized G_∞	feedforward $G_0' \beta A$
A	 $\frac{Z_2}{Z_1} + 1$	$\left(\frac{Z_2}{Z_1} + 1 \right) \left(1 + \frac{1}{X} \right) + \frac{Z_2}{Z_{cm}^-}$	$\frac{Z_0}{A} \left(\frac{1}{Z_d} + \frac{1}{X(Z_d \parallel Z_{cm}^-)} \right)$
B	 $\frac{1}{Z}$	$\frac{1}{Z} \left(1 + \frac{1}{X} \right) + \frac{1}{Z_{cm}^-}$	$-\frac{1}{A} \left(\frac{1}{Z_d} + \frac{1}{X(Z_d \parallel Z_{cm}^-)} \right)$
C	 $-Z$	$-Z$	$\frac{Z_0}{A}$

D		$-\left(\frac{Z_2}{Z_1} + 1\right)$	$-\left(\frac{Z_2}{Z_1} + 1\right)$	$-\frac{1}{A}$
E		$-\frac{Z_2}{Z_1}$	$-\frac{Z_2}{Z_1}$	$\frac{Z_0}{AZ_1}$
F		$-\frac{1}{Z_1}\left(\frac{Z_2}{Z} + 1\right)$	$-\frac{1}{Z_1}\left(\frac{Z_2}{Z} + 1\right)$	$-\frac{1}{AZ_1}$
G		1	$1 + \frac{1}{X}$	$\frac{Z_0}{A}\left(\frac{1}{Z_d} + \frac{1}{X(Z_d \parallel Z_{cm})}\right)$

TABLE IV

Closed-loop-gain static error due to inaccuracies of feedback-network components

	Operational circuit	Ideal closed-loop gain G_1	Closed-loop-gain static error ϵ_0
A		$\frac{R_2}{R_1} + 1$	$\frac{1}{1 + R_1/R_2}\left(\frac{\delta R_2}{R_2} - \frac{\delta R_1}{R_1}\right)$
B		$\frac{1}{R}$	$-\frac{\delta R}{R}$
C		$-R$	$\frac{\delta R}{R}$
D		$-\left(\frac{R_2}{R_1} + 1\right)$	$\frac{1}{1 + R_1/R_2}\left(\frac{\delta R_2}{R_2} - \frac{\delta R_1}{R_1}\right)$

E		$-\frac{R_2}{R_1}$	$\frac{\delta R_2}{R_2} - \frac{\delta R_1}{R_1}$
F		$-\frac{1}{R_1} \left(\frac{R_2}{R} + 1 \right)$	$-\frac{\delta R_1}{R_1} + \frac{1}{1 + R/R_2} \left(\frac{\delta R_2}{R_2} - \frac{\delta R}{R} \right)$
G		$-\frac{R_1}{R} \left(1 + \frac{R_2}{R_1 \parallel R_3} \right)$	$-\frac{\delta R}{R} + \frac{1}{1 + R_3/(R_1 \parallel R_2)} \times$ $\times \left[\left(1 + \frac{R_3}{R_2} \right) \frac{\delta R_1}{R_1} + \left(1 + \frac{R_3}{R_1} \right) \frac{\delta R_2}{R_2} - \frac{\delta R_3}{R_3} \right]$
H		$-\frac{1}{\omega C_2 R_1}$	$-\left(\frac{\delta R_1}{R_1} + \frac{\delta C_2}{C_2} \right)$

TABLE V
Simplified expressions for input impedance Z_i of basic operational circuits, assuming $X = \infty$, $Z_o = 0$ and $Z_u, Z_{cm}, Z_{cm}^+ \gg Z_L, Z_2, Z$

Operational circuit	Input impedance		
	ideal Z_{i1}	actual Z_i	
A		∞	$Z_{cm}^+ \parallel [Z_d(1 + \beta_{sc}A)] \approx Z_{cm}^+$ for $1/\beta_{sc} = 1 + Z_2/Z_1$
B		∞	$Z_{cm}^+ \parallel [Z_d(1 + \beta_{sc}A)] \approx Z_{cm}^+$ for $1/\beta_{sc} = 1 + Z_L/Z$
C		0	$\frac{Z}{1 + A}$
D		0	$\frac{Z_2 + Z_1 \parallel Z_L}{1 + \beta_{oc}A}$ for $1/\beta_{oc} = 1 + Z_L/Z_1$
E		Z_1	$Z_1 + \frac{Z_2}{1 + A}$
F		Z_1	$Z_1 + \frac{Z_2 + Z \parallel Z_L}{1 + \beta_{oc}A}$ for $1/\beta_{oc} = 1 + Z_L/Z$

TABLE V (continued)

Operational circuit	Input impedance	
	ideal Z_{11}	actual Z_1
G		$Z_{cm}^+ \parallel [Z_d(1 + A)] \approx Z_{cm}^+$
H		$-\frac{Z_1 Z_L}{Z_2} + \frac{Z_1(1 + Z_L/Z_2)}{1 + \beta_{oc} A}$ for $1/\beta_{oc} = 1 + Z_L/Z_2$

TABLE VI

Simplified expressions for output impedance Z_o of basic operational circuits, assuming $X = \infty$, $Z_d, Z_{cm}, Z_{em} \gg Z_o, Z_s, Z_1, Z_2, Z$ and $Z_o = 0$ (F, H)

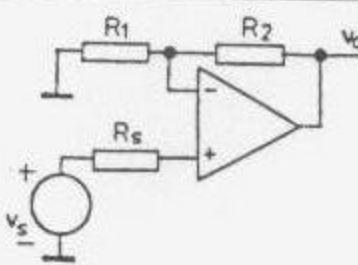
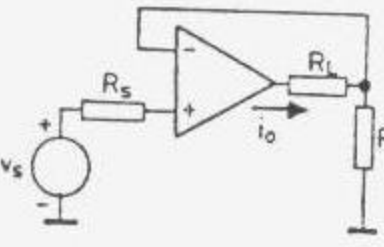
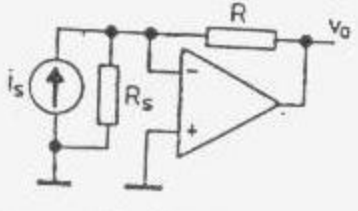
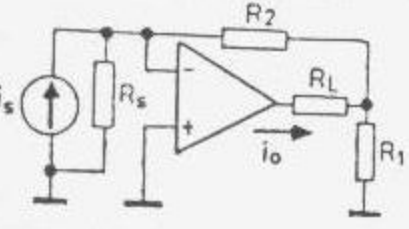
Operational circuit	Output impedance	
	ideal Z_{o1}	actual Z_o
A		$\frac{Z_o \parallel (Z_1 + Z_2)}{1 + \beta_{oc} A}$ for $1/\beta_{oc} = 1 + (Z_2 + Z_o)/Z_1$
B		$Z_o + Z(1 + A)$
C		$\frac{Z_o}{1 + A}$
D		$Z_o + Z_1(1 + A)$
E		$\frac{Z_o \parallel (Z_1 + Z_2)}{1 + \beta_{oc} A}$ for $1/\beta_{oc} = 1 + (Z_2 + Z_o)/Z_1$
F		$[Z \parallel (Z_1 + Z_2)](1 + \beta_{sc} A)$ for $1/\beta_{sc} = 1 + Z_2/Z_1$

TABLE VI (continued)

Operational circuit	Output impedance	
	ideal Z_{oi}	actual Z_o
G	0	$\frac{Z_o}{1+A}$
H	∞	$(Z_1 \parallel Z_2)(1 + \beta_{sc}A)$ for $I/\beta_{sc} = 1 + Z_2/Z_1$

TABLE VII

Input and output offset of unbalanced operational circuits

Operational circuit	Input offset	Output offset
A	 E_{osi} $-E_{os} - I_b^-(R_1 \parallel R_2) + I_b^+ R_s$	E_{oso} $E_{os} \left(\frac{R_2}{R_1} + 1 \right) + I_b^- R_2 - I_b^+ R_s \left(\frac{R_2}{R_1} + 1 \right)$
B	 E_{osi} $-E_{os} - I_b^- R + I_b^+ R_s$	I_{oso} $\frac{E_{os}}{R} + I_b^- - I_b^+ \frac{R_s}{R}$
C	 I_{osi} $\frac{E_{os}}{R \parallel R_s} + I_b^-$	E_{oso} $E_{os} \left(\frac{R}{R_s} + 1 \right) + I_b^- R$
D	 I_{osi} $\frac{E_{os}}{(R_1 + R_2) \parallel R_s} + I_b^-$	I_{oso} $\frac{E_{os}}{R_1} \left(1 + \frac{R_1 + R_2}{R_s} \right) + I_b^- \left(\frac{R_2}{R_1} + 1 \right)$

E		E_{osi} $E_{os} \left(1 + \frac{R_1}{R_2} \right) + I_b^- R_1$	E_{oso} $E_{os} \left(\frac{R_2}{R_1} + 1 \right) + I_b^- R_2$
F		E_{osi} $E_{os} \left(1 + \frac{R_1}{R_2 + R} \right) + I_b^- R_1$	I_{oso} $\frac{E_{os}}{R} \left(1 + \frac{R_2 + R}{R_1} \right) + I_b^- \left(\frac{R_2}{R} + 1 \right)$
G		E_{osi} $-E_{os} + I_b^+ R_s$	E_{oso} $E_{os} - I_b^+ R_s$
H		I_{osi} $\frac{E_{os}}{R_1 + R_2 \parallel R_3} + I_b^-$	E_{oso} $E_{os} \left(\frac{R_2}{R_3} + 1 \right) + I_b^- R_1 \left(\frac{R_2}{R_1 \parallel R_3} + 1 \right)$

I		I_{osi} $\frac{E_{os}}{(R_1 + R_2) \parallel R_s} +$ $+ I_b^- \frac{R_1}{R_1 + R_2} - I_b^+ \frac{R_2}{R_1 + R_2}$	E_{oso} $E_{os} \left(1 + \frac{R_1 + R_2}{R_s} \right) + I_b^- R_1 - I_b^+ R_2$
J		E_{osi1} $E_{os} \frac{R_1}{R_0 \parallel R_1 \parallel \dots \parallel R_m} + I_b^- R_1$	E_{oso} $E_{os} \left(\frac{R_0}{R_1 \parallel \dots \parallel R_m} + 1 \right) + I_b^- R_0$
K		dE_{osi}/dt $\frac{E_{os}}{C_1 R_2} + \frac{I_b^-}{C_1}$	E_{oso} $E_{os} + I_b^- R_2$
L		E_{osi} $E_{os} + I_b^- R_1$	E_{oso} $E_{os} \left(1 + \frac{t}{C_2 R_1} \right) + \frac{I_b^- t}{C_2}$

M		—	E_{oso} $2E_{os} + I_b^- R + \frac{I_b^- t}{C}$
N		E_{osi} $E_{os} \left(1 + \frac{R_1}{R_2}\right) + I_b^- R_1 - I_b^+ R_3$	E_{oso} $E_{os} \left(\frac{R_2}{R_1} + 1\right) + I_b^- R_2 - I_b^+ R_4$
O		E_{osi} $E_{os} \left(1 + \frac{R_1}{R_2}\right) + I_b^- R_1 - I_b^+ R_3$	I_{oso} $\frac{E_{os}}{R_3 \parallel R_4} + I_b^- \frac{R_2}{R_4} - I_b^+$
P		E_{osi} $E_{os2} - E_{os1} + I_{b1}^+ R_{s1} - I_{b2}^+ R_{s2} + (I_{b2}^- - I_{b1}^-)(R_1 \parallel R_2)$	E_{oso} $(R_2/R_1 + 1) \times$ $\times (E_{os2} - E_{os1} + I_{b1}^+ R_{s1} - I_{b2}^+ R_{s2}) + (I_{b2}^- - I_{b1}^-) R_2$

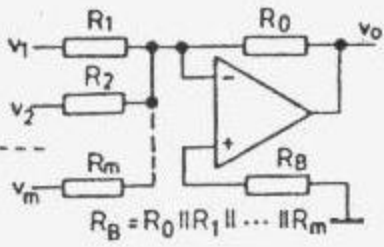
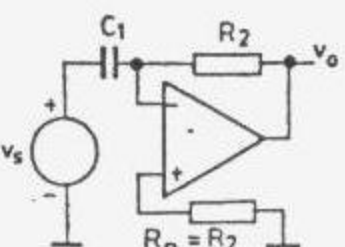
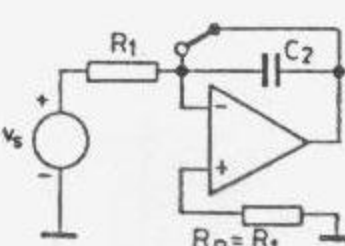
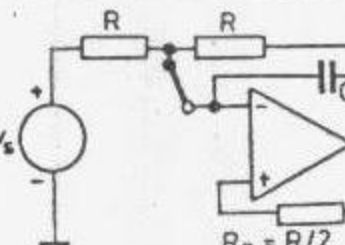
TABLE VIII

Input and output offset of balanced operational circuits

	Operational circuit	Input offset	Output offset
A1		E_{osi} $-E_{os} - I_{os}(R_1 \parallel R_2)$	E_{oso} $E_{os} \left(\frac{R_2}{R_1} + 1\right) + I_{os} R_2$
A2		E_{osi} $-E_{os} - I_{os} R_s$	E_{oso} $E_{os} \left(\frac{R_2}{R_1} + 1\right) + I_{os} R_s \left(\frac{R_2}{R_1} + 1\right)$
B1		E_{osi} $-E_{os} - I_{os} R$	I_{oso} $\frac{E_{os}}{R} + I_{os}$
B2		E_{osi} $-E_{os} - I_{os} R_s$	I_{oso} $\frac{E_{os}}{R} + I_{os} \frac{R_s}{R}$

C1		$\frac{I_{osi}}{R} + I_{os}$	$\frac{E_{oso}}{E_{os} + I_{os}R}$
C2		$\frac{I_{osi}}{R \parallel R_s} + I_{os}$	$\frac{E_{oso}}{E_{os} \left(1 + \frac{R}{R_s}\right) + I_{os}R}$
D1		$\frac{I_{osi}}{R_1 + R_2} + I_{os}$	$\frac{I_{oso}}{\frac{E_{os}}{R_1} + I_{os} \left(\frac{R_2}{R_1} + 1\right)}$
D2		$\frac{I_{osi}}{(R_1 + R_2) \parallel R_s} + I_{os}$	$\frac{I_{oso}}{\frac{E_{os}}{R_1} \left(1 + \frac{R_1 + R_2}{R_s}\right) + I_{os} \left(\frac{R_2}{R_1} + 1\right)}$

E		$\frac{E_{osi}}{E_{os} \left(1 + \frac{R_1}{R_2}\right) + I_{os}R_1}$	$\frac{E_{oso}}{E_{os} \left(\frac{R_2}{R_1} + 1\right) + I_{os}R_2}$
F		$\frac{E_{osi}}{E_{os} \left(1 + \frac{R_1}{R_2 + R}\right) + I_{os}R_1}$	$\frac{I_{oso}}{\frac{E_{os}}{R} \left(1 + \frac{R_2 + R}{R_1}\right) + I_{os} \left(\frac{R_2}{R} + 1\right)}$
G		$\frac{E_{osi}}{-E_{os} - I_{os}R_s}$	$\frac{E_{oso}}{E_{os} + I_{os}R_s}$
H		$\frac{I_{osi}}{R_1 + R_2 \parallel R_3} + I_{os}$	$\frac{E_{oso}}{E_{os} \left(\frac{R_2}{R_3} + 1\right) + I_{os}R_1 \left(\frac{R_2}{R_1 \parallel R_3} + 1\right)}$

 $R_B = R_0 \parallel R_1 \parallel \dots \parallel R_m$	E_{osi1} $E_{os} \frac{R_1}{R_0 \parallel R_1 \parallel \dots \parallel R_m} + I_{os} R_1$	E_{oso} $E_{os} \left(\frac{R_0}{R_1 \parallel \dots \parallel R_m} + 1 \right) + I_{os} R_0$
 $R_B = R_2$	$\frac{dE_{osi}}{dt}$ $\frac{E_{os}}{C_1 R_2} + \frac{I_{os}}{C_1}$	E_{oso} $E_{os} + I_{os} R_2$
 $R_B = R_1$	E_{osi} $E_{os} + I_{os} R_1$	E_{oso} $E_{os} - I_b^+ R_1 + (E_{os} + I_{os} R_1) \frac{t}{C_2 R_1}$
 $R_B = R/2$	$-$	E_{oso} $2E_{os} + I_{os} R + \frac{I_b^- t}{C}$

INDEX

- Absolute stability, 403–406
- Nyquist stability criterion, 405
- stable, unstable, conditionally stable operational circuit, 405–406
- Absolute-value circuit, 198
- Active collector load, 81–82
- Active filter, 192–193
- Active frequency band:
 - of operational amplifier, 34
 - of operational circuit, 219
- Actual band-pass filter, 379–380, 382 to 383
 - noise bandwidth, 379–380
 - signal bandwidth, 380
- Additive error, 241
- Additive parameters, 23, 25–31
- Amplitude error, 243
- Analogue switch, 186
- Aperiodic overshoot, 287–289
- Aperiodic undershoot, 284, 288–289
- Aperiodicity boundary, 288
- Auxiliary components, 251
- Average input bias current, *see* Input bias current
- Balancing, *see* Resistor balancing
- Balancing resistor, 335
- Band-gap voltage, 49
- Band-pass filter, 140, 144, 379–383
 - noise bandwidth, 379
 - signal bandwidth, 380
- Base-width modulation, *see* Early effect
- Bipolar operational amplifier, 44, 110, 116
- Bipolar transistor, 45–47, 49–51, 57–59, 61–63
 - band-gap voltage, 49
 - base-emitter voltage, 45–46
 - current dependence, 45
 - temperature coefficient, 50–51
 - temperature dependence, 50
 - voltage dependence, 61–64
 - base width, 45, 61–62
 - collector conductance, 62
 - current gain, 45
- Capacitive load, *see* Output capacitive load
- Capacitive pick-up, 399
- Capacitor, 252–254
- dielectric absorption, 253
- leakage resistance, 254
- mylar, polystyrene, teflon, 254
- series inductance, 254
- Chopper modulator, 125–129
- Chopper-stabilized operational amplifier, 110, 125–126
- Chopper-type operational amplifier, 110, 127–129
- Closed-loop, 202. *See also* Operational circuits; Types of operational circuits
- Closed-loop gain, 214–224
 - basic form, 214–216
 - calculation, 219–224
 - closed-loop gain and feedback factor, 216–217
- temperature dependence, 55
- voltage dependence, 62
- current-normalized collector conductance, 63
- current-normalized transconductance, 46 to 47
- differential emitter resistance, 47
- Early voltage, 62
- excess phase shift, 98
- noise, 55–60
- saturation current, 45, 49, 61
- super-beta, 54
- transconductance, 46
- Bistable noise, *see* Popcorn noise
- Blackman impedance relation, 311
- Boltzmann constant, 45, 56
- Booster, 440
 - current booster, 440
 - voltage booster, 440
- See also* Output stage
- Bootstrapping, 85
- Butterworth low-pass filter, 378–379
 - noise bandwidth, 379
 - signal bandwidth, 379
- Capacitive load, *see* Output capacitive load
- Capacitive pick-up, 399
- Capacitor, 252–254
- dielectric absorption, 253
- leakage resistance, 254
- mylar, polystyrene, teflon, 254
- series inductance, 254
- Chopper modulator, 125–129
- Chopper-stabilized operational amplifier, 110, 125–126
- Chopper-type operational amplifier, 110, 127–129
- Closed-loop, 202. *See also* Operational circuits; Types of operational circuits
- Closed-loop gain, 214–224
 - basic form, 214–216
 - calculation, 219–224
 - closed-loop gain and feedback factor, 216–217